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# Watermarking algorithm based on a human visual model

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## Abstract

This paper presents an additive watermarking technique for grey-scale pictures. It consists in secretly embedding copyright information (a binary code) into the picture without degrading its quality. Those bits are encoded through the phase of maximal length sequences (MLS). MLS are binary sequences with good correlation properties. The result of the autocorrelation is much greater than crosscorrelations, i.e. correlations made with shifted versions of this sequence. The embedded bits are retrieved from the result of the correlations. The core of the embedding process is underlaid by a masking criterion that guarantees the invisibility of the watermark. It is combined with an edge and texture discrimination to determine the embedding level of the MLS, whose bits are actually spread over  $32 \times 8$  pixel blocks. Eventually, some results are presented, which analyze the efficiency of the retrieval as well as the resistance of the watermark to compression and its robustness against malevolent manipulation. © 1998 Elsevier Science B.V. All rights reserved.

## Zusammenfassung

Dieser Artikel präsentiert eine additive Wasserzeichenmethode für Graustufenbilder, bei der eine verborgene Copyright-Information (ein binärer Code) ohne Beeinträchtigung der Bildqualität im Bild eingebettet wird. Diese Bits werden durch die Phase von Folgen maximaler Länge (MLS) codiert. MLS sind binäre Folgen mit guten Korrelationseigenschaften. Die Autokorrelation ist viel größer als Kreuzkorrelationen, d.h. Korrelationen mit verschobenen Versionen dieser Folge. Die eingebetteten Bits werden aus dem Ergebnis der Korrelationen wiedergewonnen. Der Kern des Einbettungsprozesses beruht auf einem Maskierungskriterium, das die Unsichtbarkeit des Wasserzeichens garantiert. Eine weitere Komponente ist eine Kanten- und Texturunterscheidung zur Bestimmung des Einbettungsniveaus der MLS, deren Bits über Blöcke von  $32 \times 8$  Pixel verteilt sind. Schließlich werden einige Ergebnisse präsentiert, welche die Effizienz der Bit-Wiedergewinnung sowie die Robustheit des Wasserzeichens gegenüber Kompression und böswilliger Manipulation untersuchen. © 1998 Elsevier Science B.V. All rights reserved.

## Résumé

Nous présentons dans cet article une technique de watermarking additif pour des images en niveaux de gris. Elle consiste à intégrer secrètement une information de droits d'auteur (un code binaire) dans l'image sans dégrader sa qualité.

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Ces bits sont encodés sur la phase de séquences de longueur maximale (MLS). Les MLS sont des séquences binaires possédant de bonnes qualités de corrélation. Le résultat de l'autocorrélation est beaucoup plus élevé que celui des intercorrélations, à savoir les corrélations obtenues avec des versions décalées de la séquence. Les bits intégrés sont récupérés à partir du résultat des corrélations. Le noyau du processus d'intégration est sous-tendu par un critère de masquage garantissant l'invisibilité du filigrane numérique (watermark). Il est combiné avec une discrimination de contour et de texture pour déterminer le niveau d'intégration de la MLS, dont les bits sont en fait disséminés sur des blocs de  $32 \times 8$  pixels. Enfin, nous présentons certains résultats qui permettent d'analyser l'efficacité du recouvrement ainsi que la résistance du watermark à la compression et sa robustesse vis-à-vis de manipulations malveillantes. © 1998 Elsevier Science B.V. All rights reserved.

*Keywords:* Copyright; Digital picture watermarking; Human vision system model; Masking; Spread spectrum

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## 1. Introduction

Copyright offers protection for the contents of the exchanges which take place on the networks. With the development of the Information Society, it is expected that more and more intellectual property and other protected material will be carried on the Superhighway. It is clear that with increasing digitization, classical legal protection is not sufficient but the development of technical protection tools will be the key to the viability of the Information Society. Indeed, the acceptance of new digital audiovisual services depends on whether suitable techniques for the protection of the content providers' interests are available [13].

As a matter of fact, its very viability is threatened by the nature of digital media. First, the replication of digital material is very easy and, more dangerous, is virtually perfect. The copy is identical to the original. The ease of transmission and multiple uses is worrying too. Once a single unauthorized copy has been made, it is instantaneously accessible to anyone who wants it, without any control by the owner of the original content. In fact, the plasticity of digital media is a great menace. Any malevolent user (*a pirate*) can modify an image at will. Such manipulations are really easy with the existing image processing tools and defy many copyright protection methods.

Fortunately, digitization of audiovisual contents offers new possibilities for the development of copyright protection techniques. Watermarking is just one. The principle of watermarking is the robust embedding of copyright information (e.g. time and date, copyright identifiers) into a given content.

This content may be a text [2,19], an audio content [1], but most often watermarking is applied to still or moving images. This paper will focus on image watermarking for still pictures (extension to moving pictures is possible by independently watermarking some of the pictures of the sequence).

Many laboratories and companies have already developed their own watermarking techniques for digital images. Most of them work directly on the luminance with or without taking considerations about the quality of the watermarked image into account [4,5]. It is also interesting to consider the image content as a channel that can convey a certain amount of information. Different techniques use this approach to embed a copyright code by means of the spread-spectrum theory [7]. Finally, other authors apply their watermarking techniques not on the picture itself but on some of its characteristics, like DCT coefficients [14], fractal coefficients or motion estimation vectors, high-resolution coefficients in case of multiresolution encoding [18], or the phase of DFT coefficients [21].

These techniques give different results with different degrees of quality, but all have in common that they have to realize a good trade-off between the robustness of the watermarking, the quality of the watermarked picture and the computational cost. However, only some analyze the quality of the resulting image or use a human visual model to guarantee the invisibility of the embedding [10]. The method presented in this paper also exploits this human visual model. Yet, it is computationally simpler. Moreover, its objective is the hiding of a secret message into the images while the other

had been conceived to answer the simple binary question: “Does the image contain a defined watermark or not?”. Of course, these two functionalities are complementary and should be combined in the automated monitoring of copyrighted material on audiovisual networks.

Another asset is the fact that the retrieval of the embedded copyright information does not require the use of the original picture, so that no human intervention is needed for the retrieval of the watermark. This advantage is essential in the use of watermarking for the automated monitoring of audiovisual networks. As a matter of fact, this monitoring permits the automatic detection of copyright violations. This is fundamental given the practical situation, where Copyright Owners possess a lot of copyrighted works and their material is distributed over and over through increasingly developing networks.

## 2. Perceptive model

The goal of image watermarking techniques is to embed some information in the picture content. In the case of an invisible watermark, the watermark must be hidden (i.e. *masked*) by the picture it is inlaid in. The question of embedding visible watermarks is completely different and is addressed by other authors [3]. The watermarking method presented in this paper precisely refers to a masking criterion deduced from physiological and psychophysical studies [6].

### 2.1. Eye functioning and masking concept

It is now admitted that the retina of the eye splits the visual stimulus composing an image in several components. These components circulate by different tuned channels from the eye to the cortex, each channel being tuned to a component. The characteristics of a component are:

- Location in the visual field (in the image).
- Spatial frequency (in the Fourier domain: the amplitude in polar coordinates).
- Orientation (in the Fourier domain: the phase in polar coordinates)

A perceptive channel can only be stimulated by a component of a signal whose characteristics are tuned to its own characteristics. Components that have different characteristics are independent. Moreover, according to the perceptive model of human vision [28], signals that have similar components use the same channels from the eye to the cortex. It appears that such signals interact and are subject to non-linear effects. *Masking* is one of those effects. It occurs when the *detection threshold*, i.e. *the minimum level below which a signal cannot be seen*, is increased because of the presence of another signal. In other words, masking occurs when a signal cannot be seen because of another signal with near characteristics and at a higher level.

### 2.2. The masking model

With the aim of making a model of the masking phenomenon, tests have been done on monochromatic signals, also called *gratings*, i.e. signals of one single frequency and one orientation  $(f_0, \theta_0)$ . It appears that the eye is sensitive to the contrast of those gratings.  $L$  being the luminance, this contrast is defined by

$$C = \frac{2(L_{\max} - L_{\min})}{L_{\max} + L_{\min}}. \quad (1)$$

The detection threshold contrast  $C_s$  of a test signal is a non-linear function of the contrast  $C_m$  of a masking signal. When the two gratings have the same frequency and orientation  $(f_0, \theta_0)$ , this threshold can be expressed as [17]

$$C_{s(f_0, \theta_0)}(C_m) = \max \left[ C_0, C_0 \left( \frac{C_m}{C_0} \right)^\varepsilon \right], \quad (2)$$

where  $C_0$  is the visibility threshold without a masking effect and  $\varepsilon$  depends on  $(f_0, \theta_0)$ , typically,  $0.6 \leq \varepsilon \leq 1.1$ .

It is possible to extend that expression to introduce the frequency and orientation dependences. Actually, the masking phenomenon decreases as the couple  $(f, \theta)$  of the test (or masked) signal differs from the couple  $(f_0, \theta_0)$  of the masking signal [24].

The general expression of the detection threshold contrast  $C_s$  becomes

$$C_s(C_m, f, \theta) = C_0 + k_{(f_0, \theta_0)}(f, \theta) [C_{s_{(f_0, \theta_0)}}(C_m) - C_0], \quad (3)$$

where

$$k_{(f_0, \theta_0)}(f, \theta) = \exp \left[ - \left( \frac{\log^2(f/f_0)}{F^2(f_0)} + \frac{(\theta - \theta_0)^2}{\Theta^2(f_0)} \right) \right]. \quad (4)$$

In this expression,  $F(f_0)$  and  $\Theta(f_0)$  are parameters that represent the spreading of the Gaussian function,  $C_0$  is often negligible. The spread of the Gaussian function depends on the frequency  $f_0$ . The spatial frequency typical bandwidths at half-response are 2.5 octaves at 1 c/d and 1.5 octaves at 16 c/d with a linear decrease between these two frequencies [29]. The orientation bandwidth at half-response depends on  $f_0$  and takes typical values like  $30^\circ$  at 1 c/d and  $15^\circ$  at 16 c/d [30].

These values fit with other physiological findings that have shown that the spatial frequency bandwidths of the cortex cells range from 0.5 to 2.5 octaves, clustering around 1.2 octaves [11,15] and 1.5 octaves [8]. In these works, the cortex cells behavior is approximated through a Gaussian filter with an elliptical envelop whose axis is parallel to the orientation feature of the cell. It has been established that the aspect ratio of the envelope is 1.5–2 [8]. It is easy to check that this is in accordance with the  $30^\circ$  half-bandwidth along  $\theta$ .

According to the same expression, the frequency dependence of the detection threshold has a Gaussian form. Only near-frequency signals can interact. When the frequency of the masking signal (the mask) is far from that of the signal to mask, the detection threshold is almost equal to  $C_0$ .

### 2.3. The masking criterion

It is important to notice that these results only concern gratings signals. To deduce a masking

criterion that applies to signals such as real images, the preceding masking condition has to be adapted. So, it is necessary to define a new concept which will be able to take the place of the contrast and is defined for real images. This new concept [6] is the *local energy* [23].

The local energy is defined on narrow band signals. *The local energy of a narrow band signal corresponds to the square of the amplitude of the signal envelope.* In the framework of the analytic representation of signals, in which signals are described only by positive frequencies, the local energy is the square modulus of the local signal complex value.

*A picture is a broadband signal.* The contrast related to a particular frequency and orientation is the local energy of the narrow band signal resulting from the filtering of the original picture by the *Gabor filter* (see Eq. (5)) whose characteristics are tuned to the perceptive component under consideration. Actually, this filter was chosen to model human perception. Its effect is close to the filtering effect of the visual cortex cells. As a conclusion, local energy is calculated according to the scheme presented in Fig. 1

Having introduced this local energy concept, *the masking criterion* can be defined:

“A noise is masked by a mask if  $\forall$  pixel  $(x, y)$  and  $\forall (f_0, \theta_0)$ ,  $E_{\text{mask}, (f_0, \theta_0)}(x, y) \geq E_{\text{noise}, (f_0, \theta_0)}(x, y)$ ”.

Of course, in practice, the Fourier frequency and phase space  $(f_0, \theta_0)$  has to be sampled. This generates a bank of filters whose central frequencies correspond to independent components spread all over the Fourier space. It is widely accepted [8,29] that 4 or 5 frequencies and 4–9 orientations are sufficient. The standard choice is 20 filters (5 frequencies and 4 orientations). Recently, Tai Sing Lee, by extending to two dimensions the frame criterion developed by Daubechies [9], has derived the conditions under which a set of continuous 2D Gabor wavelets provides a complete representation of any image [16]. He has proved that 2D Gabor

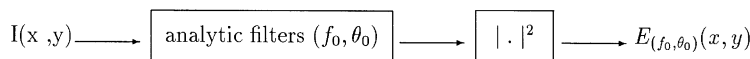


Fig. 1. Local energy computation.

wavelets with 1.5 octave bandwidth guarantee a complete representation, but only if the number of sampling orientations is above 6. Noting that the bandwidth of the filters we have used is slightly greater than 1.5 octave, this result also tends to prove that the 20 sampled filters efficiently cover the Fourier space.

#### 2.4. Perceptive filter in the horizontal direction

In the Fourier space, analytic filters used to extract the local energy are defined by

$$G_{(f_0, \theta_0)}(f, \theta) = \exp \left[ - \left( \frac{\log^2(f/f_0)}{F^2(f_0)} + \frac{(\theta - \theta_0)^2}{\Theta^2(f_0)} \right) \right]. \quad (5)$$

The embedding process presented in Section 3.1 explains that the watermark is limited to one horizontal perceptive component. So, in the following, only horizontal filters will be useful (i.e.  $\theta_0 = 0$ ). Moreover, as explained in Section 2.3, it is not necessary to verify the masking criterion  $\forall(f_0, \theta_0)$  included in this perceptive component centered around  $(f_0, \theta_0)$ . The study of the perceptive energy at  $(f_0, \theta_0)$  is sufficient.

If  $u$  and  $v$  designate horizontal and vertical frequencies (i.e. corresponding to  $x$  and  $y$  directions), if  $u_0$  is the central horizontal frequency and if the filter is narrow band according to  $u$  and  $v$  frequencies, appropriate Gabor analytic filters can be approximated by the following separable filters:

$$\begin{aligned} G_{u_0}(u, v) &= \exp \left[ - \left( \frac{\log^2(u/u_0)}{F^2(u_0)} \right) \right] \\ &\quad \times \exp \left[ - \left( \frac{v}{u_0 \Theta(u_0)} \right)^2 \right] \\ &= G_{1, u_0}(u) G_{2, u_0}(v). \end{aligned} \quad (6)$$

Making the common hypothesis that the observer is located at a distance equal to 6 times the picture height, cycles/degree frequencies can be converted into normalized ones ( $f_{\text{norm}}$ ) according to the sampling frequency. Indeed,  $N$  being the

number of columns of the picture (i.e. number of pixels in one line),

$$\begin{aligned} f_{\text{norm}} &= \left( \frac{\text{Cycles/screen}}{\text{Number of pels}} \right) \\ &= \frac{f(\text{Cycles/degree}) \times 9.53(\text{Degrees/screen})}{N(\text{Number of pels})}. \end{aligned} \quad (7)$$

As the sampling rate is the same in both horizontal and vertical directions, the normalization factor  $9.53/N$  is also valid to convert cycles/degree horizontal frequencies into normalized frequencies. So, according to the above considerations about the bandwidth of the filter defined by Eq. (4),  $F(u_0)$  and  $\Theta(u_0)$  in Eq. (6),  $u_0$  being a normalized frequency, are defined in terms of normalized frequencies as

$$F(u_0) = \frac{[(-1/15)((u_0 N/9.53) - 1) + 2.5]/2}{\sqrt{\ln(2)}}, \quad (8)$$

$$\Theta(u_0) = \frac{31 - u_0 N/9.53}{\sqrt{\ln(2)}}. \quad (9)$$

In the spatial domain  $(x, y)$ , the inverse Fourier transform of  $G_{u_0}(u, v)$  is  $g_{u_0}(x, y) = g_{1, u_0}(x)g_{2, u_0}(y)$ . For each pixel, the local energy is computed as the square of the module of the complex number obtained by convolving this filter with the picture. In Section 3.1.2, it is explained that the normalized frequency of the filters used is located around 0.15 for  $512 \times 512$  pictures. It is added that, as the reference inscription frequency is expressed in cycles/degree (8 cycles/degree), the value  $u_0 N$  is independent of picture size ( $N$ ). After generating the  $g_{1, u_0}(x)$  and  $g_{2, u_0}(y)$  filters around this frequency, it is possible to make the following statements.

- If  $g_{1, u_0}(x)$  is computed by inversion of the Discrete Fourier Transform (DFT) obtained by sampling  $G_{1, u_0}(u)$ , it is obvious that few samples are sufficient to have a good approximation of the ideal filter. Fig. 2 shows the real and imaginary part of the  $g_1$  analytic filter. It also shows, in the Fourier domain, the  $G_1$  filter resulting from a 9 samples approximation. In the following, this nine samples complex filter will be used as the horizontal part of the analytic filter.

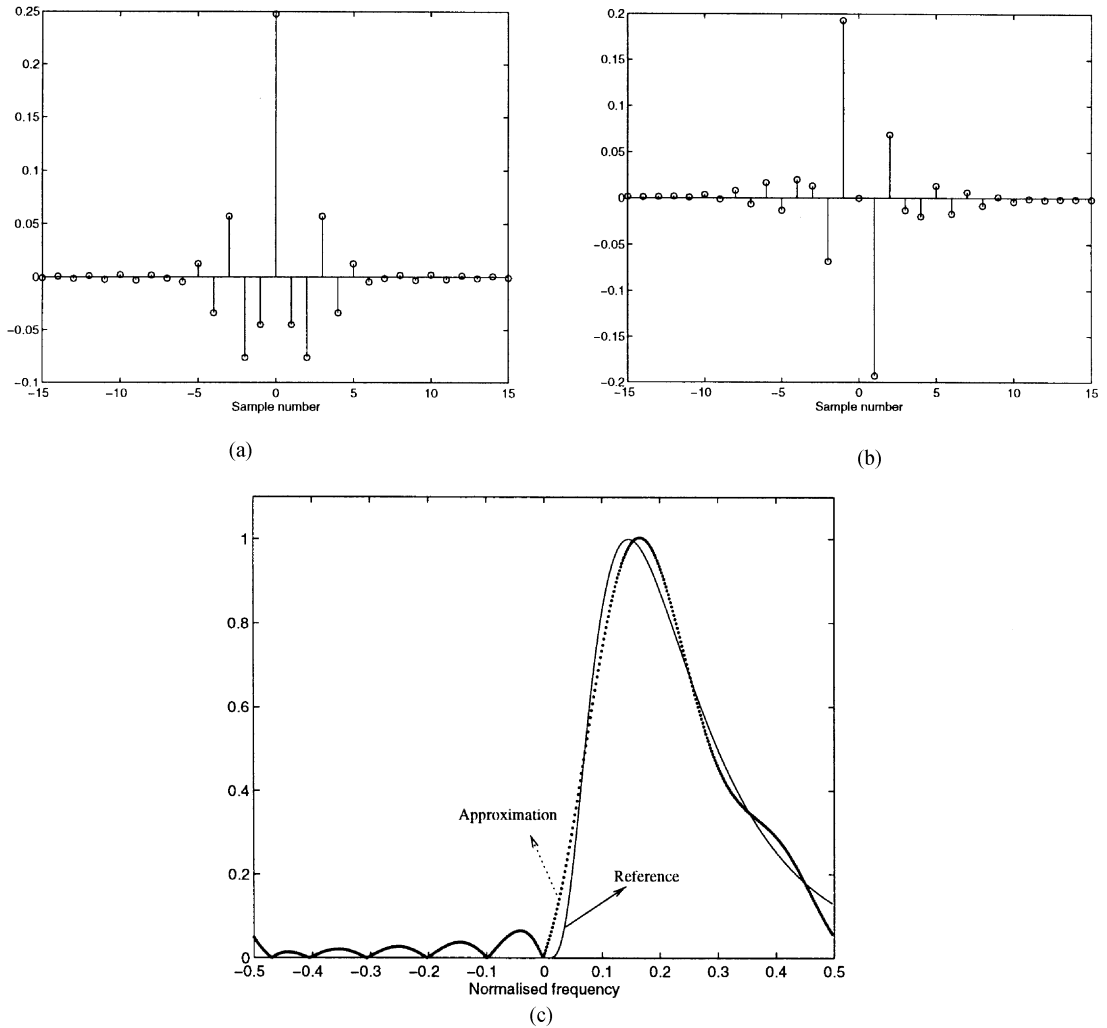


Fig. 2. Horizontal factor ( $g_1$ ) of the perceptive analytic filter ( $u_0N = 0.15N$ ): (a) real part in the spatial domain; (b) imaginary part in the spatial domain; (c) Fourier transform of the approximation using nine complex samples.

- $g_{2_{u_0}}(y)$  can be expressed analytically (as the inverse Fourier transform of a Gaussian function) (Eq. (10)). Fig. 3 shows the filter samples. In the following, the 15 most significant ones will be used to provide a genuine approximation of the filter.

$$g_{2_{u_0}}(y) = \sqrt{\pi}u_0\Theta(u_0)\exp[-(\pi y u_0\Theta(u_0))^2]. \quad (10)$$

### 3. Information embedding

#### 3.1. Main features of the method

Looking from the angle of the information theory, watermarking is similar to transmitting a bit-stream through a very noisy channel, which is the original picture, under the constraint of watermark invisibility. It is with this aim that the embedding

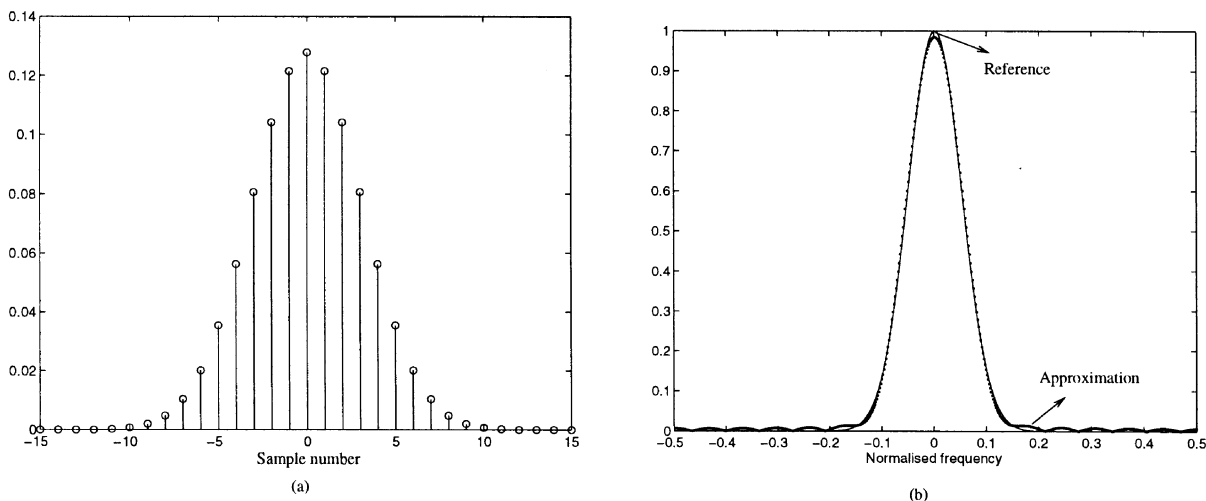


Fig. 3. Vertical factor ( $g_2$ ) of the perceptive analytic filter ( $u_0N = 0.15.N$ ): (a) filter actual samples in the spatial domain obtained as the inverse Fourier transform of the reference Gaussian function (see (b) and Eq. (8)); (b) 15 samples filter approximation.

process searches for the best trade-off while adjusting the level of the watermark. The method presented in this paper suggests adjusting this level to just below the invisibility threshold. This is done through a local examination of the picture perceptual content. Due to the heaviness of the perceptive filtering, only one component has been chosen, i.e. the horizontal component. This means that this method creates a narrow-band watermark, oriented horizontally. Only one perceptive filtering is needed to verify the masking criterion.

### 3.1.1. Maximum length sequences (MLS): *L* orthogonal codewords

As already mentioned in Section 1, the aim of the proposed algorithm is to permit embedding a secret message (identification code) into images. Interested readers are invited to read another paper [10] in which the authors describe a robust method for image authentication, i.e. an answer the key question whether image contains a defined watermark or not.

Due to the invisibility constraint, the level of the watermark has to be below that level of the picture. The picture is in fact a noise regarding the retrieval of the transmitted information. This high noise in the transmission channel leads to spread the information bitstream over binary code words. Code

words are sequences of 1 and  $-1$  symbols. The decoding of these code words is performed through correlations. This entails the use of orthogonal code words. Besides, equiprobable sequences symbols are advantageous in order to reduce the effect of the noise after correlation.

Maximal length sequences (MLS) perfectly fulfill these requirements, since MLS sequences are nearly orthogonal to their shifted versions. Crosscorrelations between shifted versions are equal to  $-1$ , whereas autocorrelations are equal to the length of the MLS sequence. They are also referred to as pseudorandom sequences, because the various statistics associated with the symbols occurrences are close to those associated with coin-toss sequences [25]. MLS exist for all integer values  $n$ , with a period  $L = 2^n - 1$ , and can be easily generated by proper connections of feedback paths in an  $n$ -stage shift register circuit [22].

In the rest of the paper, the  $L$  codewords of the  $L$ -ary system are thus shifted versions of a particular MLS. Nearly  $n$  bits are coded through the phase, i.e. the number of shifts, of an  $L = 2^n - 1$  long sequence. The choice of the length  $L$  is the result of a trade off. A long-sequence permits a more efficient and reliable decoding. On the other hand, making the sequence longer decreases the number of bits encrusted in the image. A viable compromise is the

use of a 31 symbols long sequence, i.e.  $n = 5$ . So, each time a complete sequence is embedded in the picture, nearly 5 bits of information are transmitted. As every symbol of the MLS is spread over  $8 \times 32$  pixels rectangles (see Section 3.1.2),  $512 \times 512$  pictures can contain 33 complete MLS. That means that nearly 165 bits are embedded in a  $512 \times 512$  picture. More accurately, there are 31 possible phases for each MLS, which means 31 possible messages each. This leads to a total information of  $\log_2(31^{33}) = 163$  bits.

### 3.1.2. Inscription in one perceptive component

Once the basic information has been encoded through the MLS phase, the resulting sequence has to be embedded in the picture. This has to be done in a way that ensures invisibility, undetectability, resistance towards malevolent manipulations and also guarantees adequate retrieval.

The effectiveness of masking concepts for ensuring the invisibility of extra information hidden into a picture has already been proved [10]. These masking concepts have been described in Section 2. A major feature of the human visual system is the fact that the retina of the eye splits the observed picture in components characterized by their location, spatial frequency, and orientation. In order to make a simple use of the masking concepts (see Section 3.2), it was decided to excite one single perceptive component at a time. This excited component is horizontal in our case. To generate a signal with spectral components around the central frequency of the excited perceptive channel, the MLS modulates a carrier at this frequency. This frequency changes for each MLS embedded in the picture. The choice of the frequencies is secret in order to make the watermark resistant against piracy. A malevolent user is not able to remove the watermark from the picture without knowing this secret frequency. *The choice of the possible carriers is limited to the frequencies in the interval [4,12] cycles/degree, i.e. to the interval [0.1,0.2] expressed in normalized frequencies for  $512 \times 512$  pictures.* The perceptive model is less valid and less effective for lower frequencies and the image content does not permit to watermark at a sufficient level for higher frequencies. Moreover, image features at higher frequency are less visible and can be re-

moved without altering the global picture quality. So, a filtering may easily remove the watermark.

The resulting signal has to be narrow band in both vertical and horizontal directions so as to stimulate only one perceptive component (i.e. the horizontal perceptive component). This signal is thus filtered in both horizontal and vertical directions. In order to keep the major components of the signal, each symbol of the MLS must be spread over some pixels, actually over a  $32 \times 8$  rectangle (8 lines of 32 pixels).

In the range of possible carrier frequencies [4,21] cycles/degree, c/d, the vertical half-bandwidth of the perceptive filter is about 3 c/d. According to Eq. (6), the half-bandwidth at half-response in the vertical direction is  $\sqrt{\ln(2)}\Theta(u_0)u_0$  where  $\Theta(u_0) = (31 - u_0)/(\sqrt{\ln(2)})$ . The fundamental harmonic of the watermark in the vertical direction has to be preserved after the vertical filtering limiting the watermark to a single perceptive component. If one binary symbol of the watermark is spread on  $X$  lines, this fundamental harmonic is located around  $(0.5 \times N)/(X \times 9.53)$  c/d. So,  $X$  has to be greater than  $(0.5 \times N)/(3 \times 9.53)$ . This leads to spread one symbol on 8 lines for a  $512 \times 512$  picture.

Horizontally, each symbol of the MLS is spread over 32 columns. This choice was made for retrieval efficiency purposes (see Section 3.1.1) and is sufficient to generate a narrow band watermark.

So, in the proposed implementation, each symbol of the MLS corresponds to an 8-pixels-high and 32-pixels-wide rectangle. Fig. 4(b) presents the spectrum (Fourier transform) of an embedded signal, i.e. the watermark evaluated on the well-known picture of 'Lena' (see Fig. 7(a)). The watermark appears to be spread on neighboring perceptive components. So, ideally the masking criterion should be verified for these perceptive components. Nevertheless, as the major part of the watermark is concentrated in the horizontal component, one may admit that, if the invisibility criterion is verified for this component then the watermark should be invisible in the whole picture.

As the inscription is generated on groups of eight lines, during the retrieval, each group of eight lines is also considered as a whole. Fig. 5 illustrates the receiver for this L-ary orthogonal codeword signaling. The center frequency of the band-pass filtering



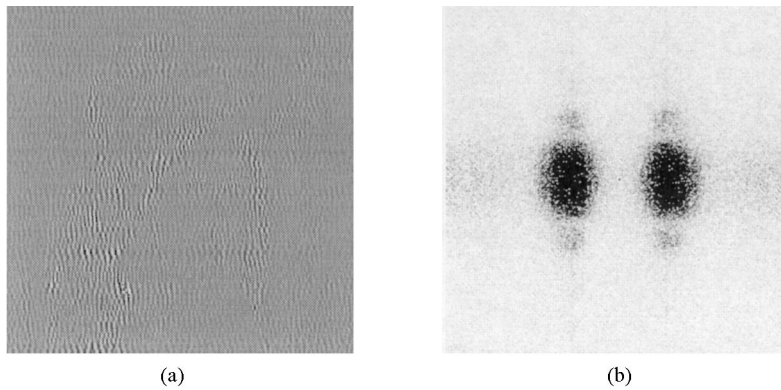


Fig. 4. Lena's watermark: (a) spatial domain, (b) Fourier space.

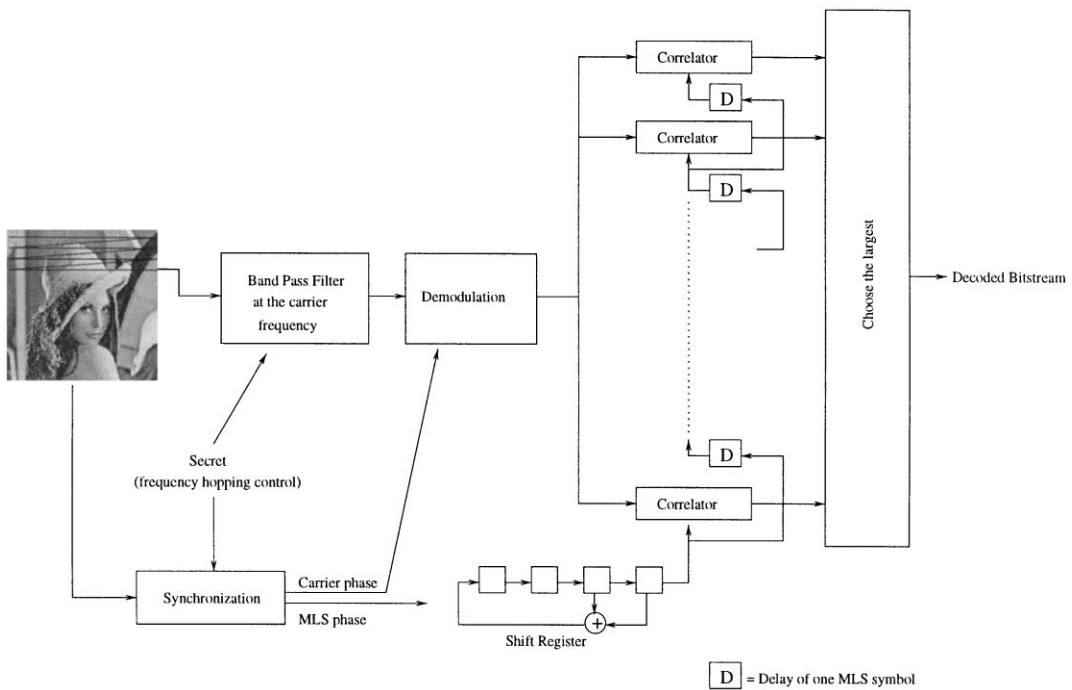


Fig. 5. Retriever for L-ary orthogonal-codeword signaling.

and the demodulation frequency are adapted to follow the frequency hopping, i.e. the change of the MLS carrier frequency. The knowledge of this frequency hopping is the secret part of the embedding process.

Synchronization is a major problem during the decoding, i.e. the acquisition of the MLS phase and

of the carrier phase. Choosing to bind the carrier phase to the MLS phase reduces these problems to one synchronization. This problem was omitted during the tests since embedding always began at the first column of the digital picture. Nevertheless, in the context of analog video product watermarking,

the digitization may be slightly different between coding and decoding. Even in the digital world some lines and columns shifting appear in the professional broadcast chain. So, this problem is crucial in a number of applications. A possible approach would be to include a pattern, typically a number of synchronization bits as part of the watermark. These bits could be located during the first stage of the retrieval through the search of a maximum correlation when computing the retrievals with different phases. Once the synchronization is achieved, the correlation properties of MLS could be used to compensate small shifting in the MLS phase, due, for example, to the loss of a column of the picture.

### 3.2. Invisibility requirement

This section describes how to determine the inscription level according to the picture content. A two-stage filtering process has been implemented:

- First, a perceptive analytic filter adapted to the frequency of the MLS carrier is used to estimate the masking capabilities of the picture in the studied area and in the perceptive component in which the MLS is embedded.
- Second, a high-frequency filter detects the main features (i.e. the edges) of the picture and, thanks to a spatial morphological filtering, distinguishes between uniform and textured areas to correct the perceptive model imperfections.

The results of these two filtering processes are then combined to produce an energy level whose square-root fixes the local inscription level.

#### 3.2.1. Perceptive analytic filtering

The masking criterion (see Section 2.3) is strongly simplified when the noise is contained in one single perceptive component. In this case, the noise designates the watermark while the mask corresponds to the original picture. The perceptive component is characterized by its horizontal orientation and by the MLS carrier frequency  $u_0$ . So, the noise – the watermark in our case – is perceptively invisible when added to the mask, i.e. the original image, if  $\forall(x, y), E_{\text{mask}, u_0}(x, y) \geq$

$E_{\text{noise}, u_0}(x, y)$  with  $E$  being the local energy defined in Section 2.3.

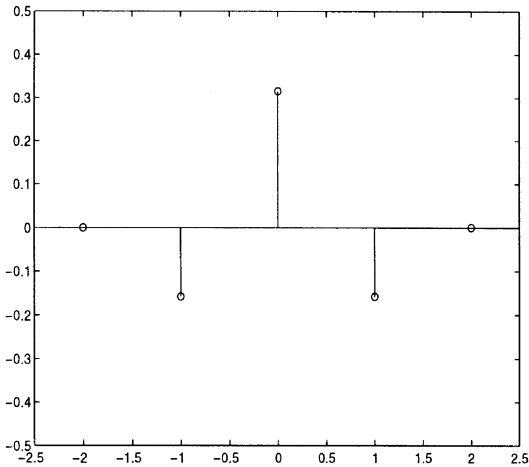
The local energy of the original image is extracted by the analytic filters defined in Section 2.4. The local energy of the image formed with the MLS could be extracted in the same way. Nevertheless, due to the small bandwidth of this signal and to the concentration of its spectrum around the central frequency of the perceptive component, it can be assumed that Gabor filtering does not modify the modulated MLS. As a result, the local energy is more or less the square of the local amplitude of the MLS signal.

So, to satisfy the masking criterion, the amplitude  $A(x, y)$  of the MLS has to be lower than the local energy of the original picture in the perceptive component centered at the frequency of the carrier modulated by the MLS. In order to produce a maximal level watermark, the amplitude is chosen equal to the local energy.

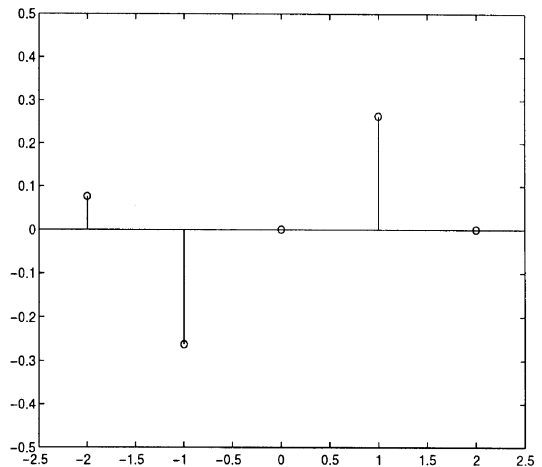
#### 3.2.2. Edges and textures discrimination

Nevertheless, this simple implementation has a number of drawbacks. In fact, the produced watermark is visible along edges and could be embedded with higher energy in textured regions, i.e. in regions that have great activity.

The problem of visibility reveals some of the weaknesses of the perceptive model. Because of their width (15 pixels), perceptive filters spread the peak of energy near sharp edges on some pixels, leading to the embedding of the watermark in a rather large area around the edges. Specific studies about the masking phenomenon around sharp edges have shown that this phenomenon is strictly local, i.e. it concerns only a very few pixels around the edge [20]. A way out of this problem is to make use of *edge detector filters*. These filters have good spatial location, i.e. they permit to precisely localize sharp edges. The used edge detectors evaluate the envelope of the original signal filtered by a large bandwidth complex filter [26] in the way described in Section 2.3, thus it can be considered another kind of local energy. In one dimension, impulse response of the analytic representation of this filter is illustrated in Fig. 6 (it is worth noting that the maximal value of the Fourier transform of this filter is 2 and not 1 as for the perceptive filters).



(a)



(b)

Fig. 6. (a) Real and (b) imaginary part of the samples of the edge detector filter.



(a)



(b)

Fig. 7. Edge detector filters applied on Lena. (a) Original grey-scale picture. (b) Edge detector local energy determined by edge detection (darkness represents high values).

This complex response is  $S + jA$  with

$$S = [0, -0.158, 0.315, -0.158, 0], \tag{11}$$

$$A = [0, -0.158, 0.315, -0.158, 0]. \tag{12}$$

In two dimensions, the energy results from the square of the sum of the energy estimated in horizontal and vertical directions. Fig. 7 presents the

local energy produced by these edge detector filters. Nevertheless, such filters do not discriminate isolated edges from edges in active areas. This is why they are followed by a non linear *morphological filtering* [12]. A *closing* of the edges detector energy increases the energy level between neighboring edges but does not modify it in the surroundings of isolated edges. The *structurant element size* fixes

what neighboring edges are. The resulting energy has strong values located on sharp edges and spread on active areas, i.e. areas having a strong edges density.

In conclusion, the information produced through this edge detector and morphological filtering brings precise information about the location of the picture masking features. This permits to correct the result obtained from the perceptive filtering, which provides a direct information about the level of inscription allowed in the perceptive component of the carrier frequency but suffers from weak location capabilities. Actually, the MLS amplitude corresponds to the minimum between the perceptive energy and twice the edge detectors energy value (it is worth noticing that both perceptive and edges detector filters have been normalized, i.e. their maximal value in the frequency domain is equal to one). This correction process does not heavily increase the computational cost but permits to reduce the embedding level near isolated edges. The resulting embedding process scheme is illustrated in Fig. 8. The watermarked bitstream contains some synchronization bits (see end of Section 3.1.2). Some parts of this figure need further explanation. The lowest branch of the scheme evaluates the sign of the original picture contribution in the correla-

tion computation. The sign of the embedded MLS is chosen to increase the absolute value of this contribution. The last band pass filter, coming before the addition of the watermark to the original picture, ensures that the embedded signal is really narrow band (see Section 3.1.2). Finally, the limiter maintains pixel values between 0 and 255.

## 4. Results

### 4.1. Invisibility and information decoding efficiency

Fig. 9 compares original and watermarked pictures having different content features. The resulting images have an acceptable quality, the pictures of Demi and Lena, in particular. However, the second picture (Boat) has lost some quality after embedding, especially near the edges (the masts). There is a wave effect in these areas that is visible on a good quality display or after high-quality printing. This is the worst case for this method, which is still not perfect for embedding near sharp edges. Corrections brought about by edge detection and morphology still has to be improved in the future. Anyway, the quality of the watermarked picture is far from disastrous.

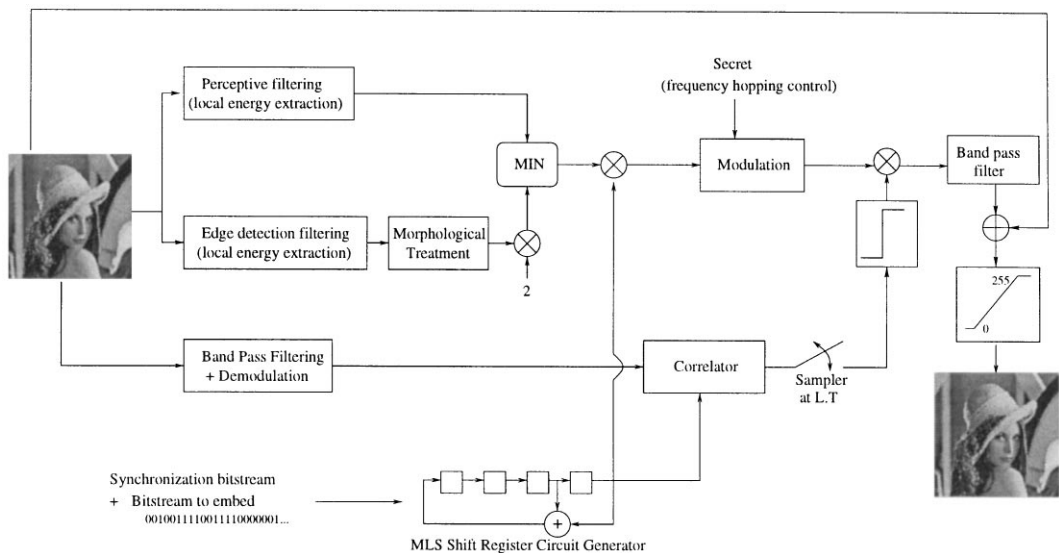


Fig. 8. Watermark embedding process.



(a.1)



(a.2)



(b.1)



(b.2)



(c.1)

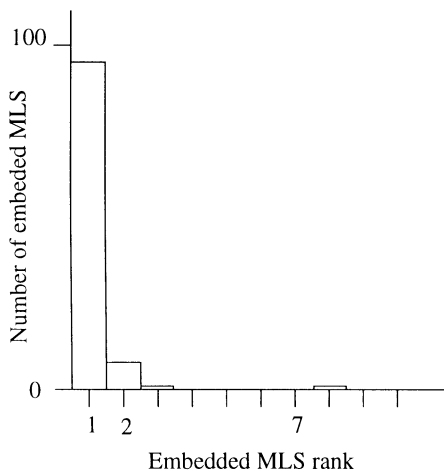


(c.2)

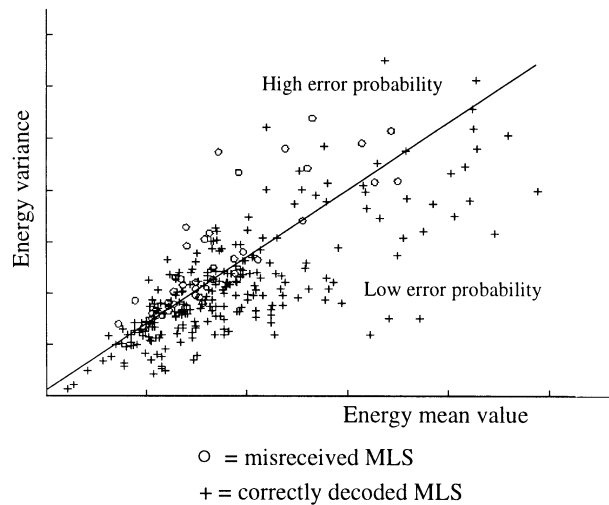
Fig. 9. Invisibility verification for (a) Lena, (b) Boat, (c) Demi: (1) original picture, (2) watermarked picture.

Picture	Lena	Boat	Demi	Average
Bit Error Rate (%) (Misreceived sequences number)	1.8	3.0	2.4	2.6

(a)



(b)



(c)

Fig. 10. Decision step effectiveness: (a) error rate (= number of MLS incorrectly decoded); (b) histogram of the rank of the embedded MLS reception in the decreasing ordered list of the reception values generated by all the shifted versions of the MLS; (c) errors occurrence in an energy variance versus energy mean graph.

This analysis of retrieval tests revealed that the MLS phase is incorrect in only a few cases (Fig. 10(a)). Moreover, some a priori knowledge may help to predict errors occurrence, since these mainly occur in regions with a high-energy variance and energy mean value ratio (Fig. 10(c)). An effective error detecting/correcting strategy has not been developed yet but it may be useful to locate an error once it has been detected, e.g. through error detecting codes [24]. Finally, Fig. 10(b), shows that when an error occurs, i.e. there exists a shifted MLS giving a greater correlation value than the one obtained by correlation with the really embedded MLS, the correct phase may be recovered simply by taking the sequence having the second correlation value. The addition of a simple BCH code

would still allow to keep most of the 163 embedded bits for the actual copyright information.

The watermarking method has also been applied on a database of 40 images having very different content features. These images are ranging from very textured images to images containing a lot of edges. The average retrieval bit error rate is shown in the last column of (Fig. 10(a)).

## 4.2. System robustness

### 4.2.1. Noise

Resistance to white noise is a minimal requirement that both methods fulfill. This is due to the correlative approach for the retrieval. The noise,



Fig. 11. (a) Lena with addition of white noise, (b) Lena blurred with a  $7 \times 7$  filter.

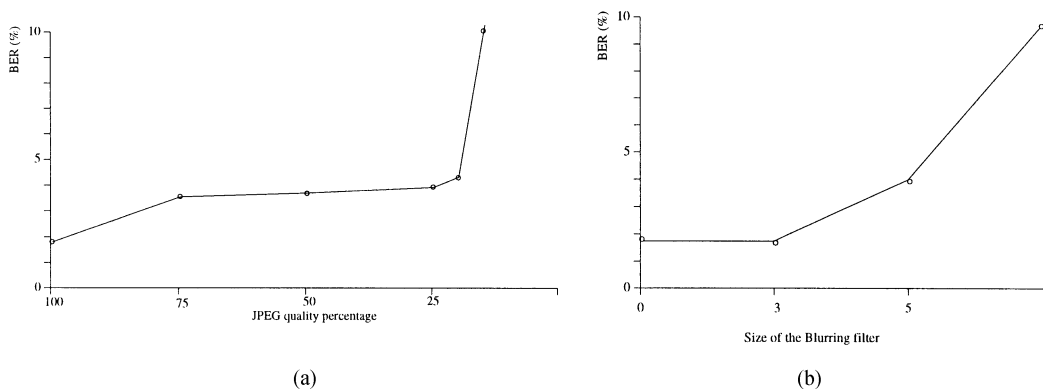


Fig. 12. (a) Evolution of the BER regarding the JPEG quality percentage, (b) evolution of the BER regarding the size of the blurring filter.

whose variance is 100 dB has nearly no influence on the detection. The BER for the retrieval tried on Lena has the same value as the retrieval without noise, i.e. 1.8%. Fig. 11(a) is the picture of Lena altered by the addition of white noise (variance 100 dB).

#### 4.2.2. JPEG [27] and low-pass filtering

Resistance against lossy compression is essential to a watermarking method. Actually, a great part of the digital pictures that are exchanged on digital networks are compressed. What is more, it would be very simple to erase the watermark if it is not resistant against compression. The method presented in this paper reacts well to compression. Fig. 12(a) shows the evolution of the bit-error-rate after

retrieval regarding the JPEG quality percentage for the picture of Lena and for the images of the database used for the results of Section 4.1. It appears that the compression does not significantly impede the retrieval, provided that the quality of the compressed picture is sufficient, i.e. above 20%. These positive results are not surprising because of the frequency range of the watermark. These frequencies are generally important for the original picture, whereas the compression algorithm removes the useless information. After comparing the results of the retrieval with and without compression, it seems that the error correcting code will allow dealing with compression without significantly diminishing the embedded bits of

information. The same conclusion applies for resistance against low-pass filtering, since the energy of the watermark is essentially concentrated in a frequency range where the image content is important. Fig. 12(b) illustrates the resistance against  $3 \times 3$ ,  $5 \times 5$  and  $7 \times 7$  blurring filters. Fig. 11(b) is the picture of Lena altered by a  $7 \times 7$  blurring filter.

#### 4.2.3. Resistance against forgery

The most definitive attack consists of replacing the embedded watermark by a new one in favor of the pirate. This section analyzes the total number of possibilities available for the embedding. It is essential to robustness against exhaustive attacks and overwatermarking. This is very important. It must be possible, in practice, to permit more than one watermark in the same image, without altering the previous ones too significantly. With this aim in mind, there must be enough independent positions for the watermark. The parameters of the embedding process include the choice of the MLS and the carrier frequency. As far as the carrier frequency is concerned, about 100 frequencies are available in the interval  $[0.1, 0.2]$  of the normalized frequencies, because carriers having a  $\Delta f_{\text{normalized}} \geq 0.1/100$  can be considered orthogonal on the length of an MLS. Moreover, for each embedded MLS, one can choose a different carrier. There are 33 embedded MLS. There are thus  $100^{33}$ , i.e.  $2^{220}$  possibilities to choose the embedding parameters. This number is huge but does not represent the actual system robustness against a pirate because the analysis of the decoding correlations obtained by decoders tuned on 100 distinct frequencies and the 186 possible MLS may convey information about the parameters used due to the appearance of particular peak features in the correlation. Nevertheless, due to his ignorance of the synchronization bits, the pirate is confronted with a strong phase acquisition problem: “How to choose the phase of the MLS carrier during the retrieval process?”. It is noticeable that the very high-noisy channel features prevents the pirate from using a classical *phase locking loop*. The same positive conclusion can be made for overwatermarking. If someone adds another watermark without any a priori knowledge of the frequencies used for the first one, he will use the same frequencies for 1% of the MLS on average. When

this occurs, half the MLS bits will be altered, the other half will be more strongly embedded. This will produce an average percentage of 0.5 errors on the whole image during the retrieval.

## 5. Conclusion

The main asset of the method presented in this paper is the use of a visual perceptive model in the production of an invisible watermark. The results are encouraging, even if in a few cases some drawbacks in the embedding method and the model itself appear. Further research would most probably solve these occasional problems. Moreover, this method reveals some interesting properties, such as the possibility of synchronization. The results seem encouraging. The bit-error-rate after retrieval is very low, so that an adapted error-correcting code would permit keeping a great part of the 163 embedded bits for copyright information. The resistance against lossy compression and overwatermarking are other very good results and can easily be dealt with by the above mentioned error correcting code.

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